

## Design a PI Controller Based on PSO and GWO for a Brushless DC Motor

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### ABSTRACT

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*brushless direct current (BLDC) motor, proportional-integral (PI) controller, particle swarm optimization (PSO), grey wolf optimization (GWO)*

Recently all the moving mechanical parts that are subjected to wear and cause errors in the future are replaced with the equivalent of electrical. A Brushless Direct Current (BLDC) motor is preferable compared to a brushed DC motor because it substitutes the unit of mechanical commutations with an electronic unit, enhancing dynamic properties, noise level, and efficiency. Since it is fairly inexpensive, simple in structure, and performs well, maximum BLDC motor drives use a Proportional-Integral PI controller for controlling the machine's speed. The major issue with the PI controller, on the other hand, is altering its parameters throughout the deployment. As a result, this work shows how to tune the PI controller settings of a BLDC motor drive using Grey Wolf Optimization (GWO) and Particle Swarm Optimization (PSO). The results of a comparison of PSO and GWO for BLDC motors were obtained. Simulation tests for the BLDC engine in MATLAB/Simulink environment show that both PSO and GWO of BLDC motor give good results, but the best is GWO in tested in terms of transient response under different mechanical loads and speeds.

## 1. INTRODUCTION

Because of their superior qualities and performance, brushless direct current motors (BLDC) have quickly acquired popularity and are now widely employed in a variety of consumer and industrial applications. A few of the advantages of BLDC motors over standard DC motors are excellent dynamic response, improved efficiency, enhanced speed versus torque properties, long-life operation, high-speed ranges, less electromagnetic interference, less noise operation, good heat dissipation, and compact size. Robotics, computer peripherals, actuation drives, machine tools, electric propulsion, and electrical power generation are all popular uses for BLDC motors [1, 2]. However, the main disadvantages are the higher cost, the size of the motor, and the requirement for a unique mounting solution for the sensors. Furthermore, Hall sensors are temperature sensitive, limiting the motor's operation, which could degrade system reliability due to the additional components and wiring [3-5]. Many researchers have used different approaches in the control speed of brushless dc motor (BLDC). In 2015, Venkata et al., [4] introduced a hybrid fuzzy PI controller for the BLDC motor speed control in electric vehicles (EV). The Hybrid Fuzzy PI controller is a combination of 2 controllers: classic PI and fuzzy PI controllers, with a selection switch which is dependent upon speed error. In the case when the speed error is over 10% of reference speed, the switch selects the FLC; or else, it selects the classic PI controller. The results reveal that the BLDC motor responds better to the Hybrid Fuzzy PI controller. Then, Yasien and Mahmood [5] presented a comparison between the new control system (NCS) and the conventional control system (CCS) of the BLDC motor. In addition, the simulation tests for BLDC motor in a MATLAB/Simulink environment show that the NCS of

BLDC motor is considered better than CCS in tested in terms of transient response under different mechanical loads and speeds. After that, Aymen et al. [6] using PSO, an important metaheuristic optimization search method, this work provides a design regarding an optimal PI controller for the BLDC motor speed control.

A TMS320F28335 DSP board with MATLAB/SIMULINK interface is used in the proposed control system. Dutta and Nayak [7] through replacing a mechanical commutation unit with an electronic unit, a BLDC motor surpasses a brushed DC motor in terms of dynamic characteristics, noise reduction, and efficiency. Since it is reasonably inexpensive, simple to develop, and performs well, maximal BLDC motor drives utilize a PID controller for controlling the machine's speed. The biggest issue with PID controllers, however, is altering their parameters during deployment. Recent research has found that the Particle Swarm Optimization (PSO) approach performs well when optimizing PID controllers. The GWO algorithm is described in this work, and it is utilized for optimizing the parameters of the PID controller. The purpose of the present work is to compare the results of tuning a PID controller with the use of PSO and GWO approaches, and to conclude that the recommended strategy delivers better dynamic performance for BLDC motors. Lastly, utilizing the Artificial Bee Colony (ABC) algorithm, Vanchinatan and Selvaganesan [8] introduced adaptive Fractional Order PID (FOPID) controller for increasing the performance of BLDC motors. From previous studies, it was seen the BLDC motors demanded their high performance and reliable speed regulation under a variety of load and speed conditions.

The aim of this research is how to design a PI controller to regulate and control the speed of controlling the brushless DC motor (BLDC) and to obtain the required amplitude value for the voltage generated by the inverter, so we proposed the two

famous methods of BLDC: Particle swarm optimization (PSO) and Gray Wolf Optimization (GWO) So that we know which is faster to reach an optimal value to PI Controller gains. Found that the speed error The brushless DC motor is optimized by 3% in PSO and 2% in GWO, respectively. Therefore, the results are considered satisfactory and it is proved that GWO is better than PSO.

## 2. PROBLEM STATEMENT

To find the best values of  $K_p$  and  $K_i$ , two solutions have been proposed to address the response speed, the first is the PSO and GWO algorithm, both of which depend on the Integral Square Error (ISE) function as a target function to calculate the error rate, and based on it, the values of  $K_p$  and  $K_i$  were determined.

## 3. METHODOLOGY AND MODELLING

This section presents the method of the work and modelling of brushless dc motor (BLDC), Modelling is classified as:

### 3.1 BLDC motor's mathematical model

The wave-form of the applied current is rectangular, while the back electromotive force (EMF) produced by the BLDCM is trapezoidal. We'll use a 3-phase, 6-state BLDCM with Y-connected windings as well as 2-phase excitation as an example in order to make things easier. We make the next assumptions, to the extent possible: i) 3-phase windings are symmetric, ii) magnetic saturation has been ignored, iii) the hysteresis and the eddy current loss values are ignored, iv) each motor winding's inherent resistance is  $R$ , self-inductance has been represented by  $L$ , and the mutual inductance has been represented by  $M$ . Thus, the 3-phase stator voltage balancing equation could be defined using state equation below [9] Figure 1 shows the dynamic equivalent circuit of the BLDC Motor.

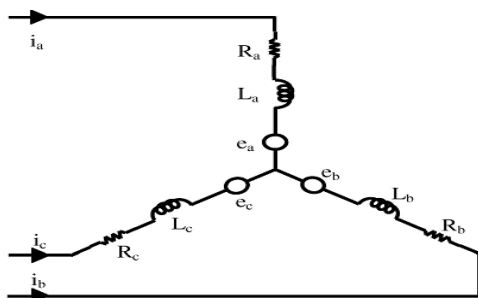


Figure 1. BLDC motor equivalent circuit

$$\begin{bmatrix} V_a \\ V_b \\ V_c \end{bmatrix} = \begin{bmatrix} R_s & 0 & 0 \\ 0 & R_s & 0 \\ 0 & 0 & R_s \end{bmatrix} \begin{bmatrix} I_a \\ I_b \\ I_c \end{bmatrix} + \begin{bmatrix} L_a & L_{ab} & L_{ac} \\ L_{ba} & L_b & L_{bc} \\ L_{ca} & L_{cb} & L_c \end{bmatrix} \frac{d}{dt} \begin{bmatrix} I_a \\ I_b \\ I_c \end{bmatrix} + \begin{bmatrix} e_a \\ e_b \\ e_c \end{bmatrix} \quad (1)$$

$L_a=L_b=L_c=L$   
 $L=L=L=M=0$

The induced backs EMFs have a trapezoidal form and their peak values are equal to  $\lambda m w$ . Electromagnetic torque could be calculated through [10]:

$$T_e = e_a I_a + e_b I_b + e_c I_c / w = \lambda m (f_a I_a + f_b I_b + f_c I_c) \quad (2)$$

where,  $f_a$ ,  $f_b$ , and  $f_c$  have shapes like  $e_a$ ,  $e_b$ , and  $e_c$ , respectively, and their maximal values are one. For a simple system with inertia  $J_s$ , load torque  $T_L$ , and friction coefficient  $B_s$ , the equation of motion is expressed as [9]:

$$J_s \frac{dw}{dt} + B_s w = T_e - T_L \quad (3)$$

The rotor speed ( $w$ ) and the rotor position ( $\theta$ ), can be written as [9]:

$$\frac{d\theta}{dt} = \frac{P}{2} w \quad (4)$$

### 3.2 BLDC motor drive's control system

The three basic components of the BLDC motor drive control system are the BLDC motor, three-phase inverter, and control system. The block diagram design regarding the control system scheme for the BLDC motor is shown in Figure 2.

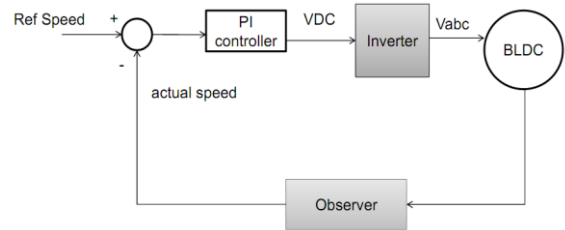


Figure 2. Diagram of a control system for BLDC motor drive

### 3.3 PI speed controllers

As a result of its low cost, ease of use, and capacity to be employed in various applications, the proportional-integral (PI) controller is widely used in manufacturing. It also improves the system's dynamic responsiveness and eliminates or decreases the error susceptibility and steady-state error [11]. In closed-loop control, a PI controller responds to an error signal that specifies the difference between the actual and desired signals by modifying the regulated value until the needed system response is reached [12]. Proportional integral controller (PI) contains two components of proportionate gain ( $K_p$ ) and integral gain ( $K_i$ ) parts. The PI controller has the general form [11]:

$$u(t) = K_p e(t) + \int_0^t K_i e(t) dt \quad (5)$$

where,  $u(t)$  represents the output of the PI controller and  $e(t)$  represents the error signal. Figure 3 shows a Proportional-Integral PI controller.

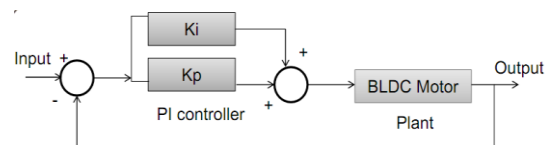


Figure 3. Block diagram representation of PI controller

## 4. OPTIMIZATION ALGORITHMS

### 4.1 Particle swarm optimization (PSO)

Eberhart and Kennedy first created the PSO technique in 1995 as a population-based optimization approach and a form of evolutionary computation tool. The technique was shown to be robust in addressing problems with nonlinearity and no differentiability, high dimensionality, and several optima, using adaptation, which is based on social-psychological theory. The next is the technique's properties: The approach is straightforward to implement, with great computational efficiency and a stable convergence characteristic, and it depends on swarm research, like bird flocking and fish schooling. Particles are a population of possible solutions that the PSO flow through search space. In the PSO method, particles have a variable velocity which affects how far they travel in search space. Every one of the particles has a memory as well, which allows it to recall the optimal solution in every one of the search spaces it has visited. In addition, Pbest refers to the place with the best fitness, whereas gbest refers to the overall best of the population's particles. In this work [13], PSO was utilized for tuning the weight coefficients in order to obtain an optimal controller. In PSO, a swarm of randomly selected individuals referred to as particles is produced. Each one of the particles represents a possible solution to the problem of optimization. Figure 4 shows the mathematical model is built using the following data [14].

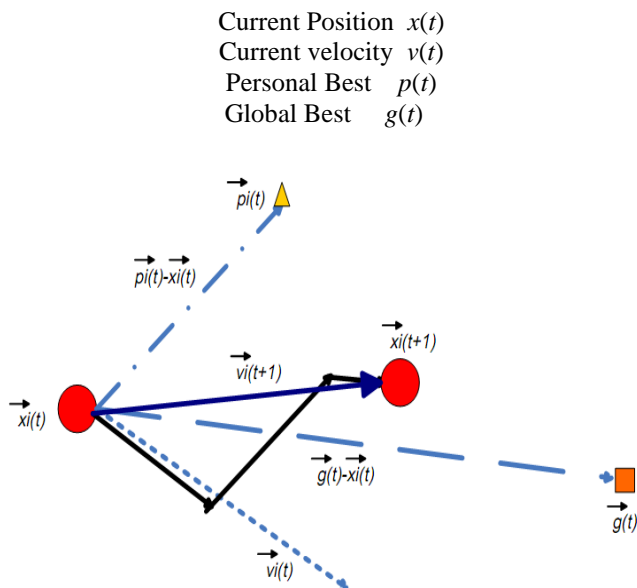


Figure 4. A mathematical model of PSO

Pseudo-code of PSO is described in Algorithm 1.

#### Algorithm 1 PSO

Step 1: Reading data and randomly generating an initial solution.

$$x_{i,j} = (x_{1,1}, x_{1,2}, x_{1,3}, \dots, x_{pop,n}),$$

$$i=1 \text{ to } pop \text{ and } j=1 \text{ to } n$$

$$v_{i,j} = (v_{1,1}, v_{1,2}, v_{1,3}, \dots, v_{pop,n}),$$

$$i=1 \text{ to } pop \text{ and } j=1 \text{ to } n$$

where, pop represents the size of the population and problem dimension.

Step 2: Calculating the objective function's fitness value.

Step 3: Calculating the pbest, which represents the value of the objective function of every one of the particles in the current iteration's population is compared to the preceding iteration and the particle position that has a lower objective function value as pbest for the Current iteration has been stated [15]:

$$pbestm^{k+1} = \{pbestm^k \text{ if } fm^{k+1} \geq fm^k\}$$

$$= \{xm^{k+1} \text{ if } fm^{k+1} \leq fm^k\} \quad (6)$$

where,  $k$  represented the number of iterations, and  $f$  represents the objective function that has been assessed for the particle.

Step 4: Calculating the value of gbest which represents the optimal objective function that is related to pbest amongst all of the particles in that iteration is compared to that in the preceding iteration and the lower value has been chosen as the current overall gbest.

$$gbestm^{k+1} = \{gbestm^k \text{ if } fm^{k+1} \geq fm^k\}$$

$$= \{pbestm^{k+1} \text{ if } fm^{k+1} \leq fm^k\} \quad (7)$$

Step 5: Velocity update, after calculating gbest and pbest, the particles' velocity for the following iteration must be updated with the use of equation [15]:

$$vm^{k+1} = \omega vm^k + C1.rand1 \times (pbesti - xm^k)$$

$$+ C2.rand2 \times (gbestm - xm^k) \quad (8)$$

where, the above equation's parameters have been specified in advance and  $\omega$  represents the weight factor of the inertia, which has been characterized as [13]:

$$w = wmax - \frac{(wmax - wmin)}{itermax} * iter \quad (9)$$

$C1$  and  $C2$  represent the coefficients of acceleration typically ranging within  $[1, 2]$  A large weight of the inertia ( $w$ ) facilitates global search whereas the small weight facilitates the local search.

Step 6: Checking the constraints of velocity components that occur in limits from conditions below:

$$\text{if } vid > vmax \text{ then } vid = vmax(\text{or}) \text{ if } vid < -vmax \text{ then } vid = -vmax$$

Step 7: update the position, where every particle's position at the following iteration ( $k+1$ ) is updated as:

$$xj^{k+1} = xj^k + vj^{k+1} \quad (10)$$

Step 8: In the case where a number of the iterations hits the maximal value, then go to step9 for the gbest, or else, go to step 2.

Step 9: The individual generating the latest value of the gbest is optimum PI parameters at minimal objective function. The diagram for implementing the PSO algorithm has been illustrated in Figure 5.

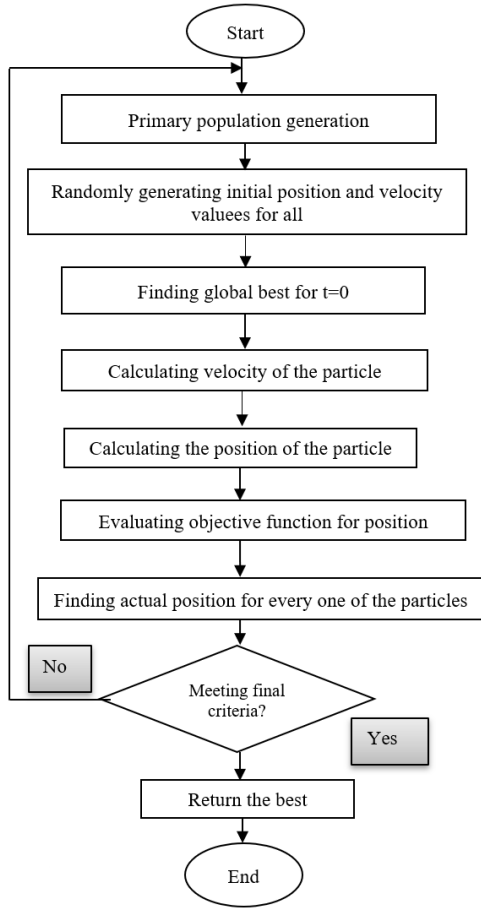


Figure 5. The diagram for implementing the PSO algorithm

#### 4.2 Grey wolf optimization (GWO)

It's a metaheuristic search method inspired by grey wolf behaviour. This algorithm has been invented by (S. Mirjalili) in 2014, and its accuracy was good when compared to classic algorithms like GA, Differential calculus, PSO, and SI. The wolves in this algorithm have been placed in a pyramid shape, with levels as illustrated in Figure 6. Alpha wolves are at top of the food chain and are the group's leader, followed by the remainder of the flock. It is believed to be a helper (alpha) in decision-making at the second level, which is known as (beta). In the third level, the (Delta) sends out wolf scouts. The final level is Omega, which is for wolves who follow the dictates of the dominating wolves. Grey wolves track, chase, encircle, and attack animals as part of their hunting habit. Alpha wolves are the ones who conduct the majority of the hunting. Because these wolves have a strong understanding of prey spots, the beta may start but is directed by the alpha. As a result, three options are most appropriate for the problem. Omega wolves use alpha, beta, and delta sites to update their sites. The first ideal solution is alpha, while 2<sup>nd</sup> and 3<sup>rd</sup> ideal solutions are respectively (beta) and (delta) [7-16].

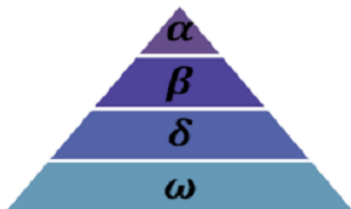


Figure 6. Hierarchy of grey wolf

In this algorithm, 4 groups have been defined, similar to the social hierarchy of the grey wolves (to live in packs), which are: Alpha ( $\alpha$ ), Beta ( $\beta$ ), Delta ( $\delta$ ), and Omega ( $\omega$ ). Wolves' social hierarchy has been modelled at the stage of the design. Alpha is the optimal solution, followed by  $\beta$  and  $\delta$  as 2<sup>nd</sup> and 3<sup>rd</sup> best options. The remaining solutions are regarded as Omega since they are the least important [17].

##### 4.2.1 Searching for prey

The grey wolves hunt for prey based on positions of the and  $\delta$ . They split apart to look for the prey.

$$|A| > 1$$

##### 4.2.2 Encircling the prey

The following equations are given to quantitatively model the encircling behaviour [17]:

$$\vec{D} = |\vec{C} \cdot \vec{Xp}(t) - \vec{X}(t)| \quad (11)$$

$$\vec{X}(l+1) = \vec{Xp}(t) - \vec{AX} \cdot \vec{DX} \quad (12)$$

where,

$$\vec{A} = 2\vec{a} \cdot \vec{r1} - \vec{a} \quad (13)$$

$$\vec{C} = 2 \cdot \vec{r1} \quad (14)$$

$$\vec{a} = 2 * \left(1 - \frac{l}{l_{max}}\right) \quad (15)$$

$l$  = Current iteration

where,  $l$  is the Current iteration.  $\vec{a}$  components are reduced linearly from 2 to 0 during iterations  $\vec{r1}$  &  $\vec{r2}$  represent random vectors in [0,1]. The grey wolves can encircle victims. The mathematical model assumes that the prey does not know its location. As a result, alpha, beta, and delta have more understanding concerning prey's location. The three best candidate answers are alpha (the first best solution), beta, and delta. Omega wolves reposition themselves by the upper layer wolves. In this approach, the following equations, namely themselves by the upper layer wolves. Figure 7 shows the flowchart of GWO method. In this approach, Eqns. (16), (17), and (18) are proposed [17]:

$$\begin{aligned} \vec{D}\alpha &= |\vec{C1}\vec{X} \cdot \vec{X}\alpha - \vec{X}| \\ \vec{D}\beta &= |\vec{C2} \cdot \vec{X}\beta - \vec{X}| \\ \vec{D}\delta &= |\vec{C3} \cdot \vec{X}\delta - \vec{X}| \end{aligned} \quad (16)$$

$$\begin{aligned} \vec{X1} &= \vec{X}\alpha - \vec{A1} \cdot \vec{D}\alpha \\ \vec{X2} &= \vec{X}\beta - \vec{A2} \cdot \vec{D}\beta \\ \vec{X3} &= \vec{X}\delta - \vec{A3} \cdot \vec{D}\delta \end{aligned} \quad (17)$$

$$\vec{X}(l+1) = \frac{\vec{X1} + \vec{X2} + \vec{X3}}{3} \quad (18)$$

##### 4.2.3 The mathematical model for hunting

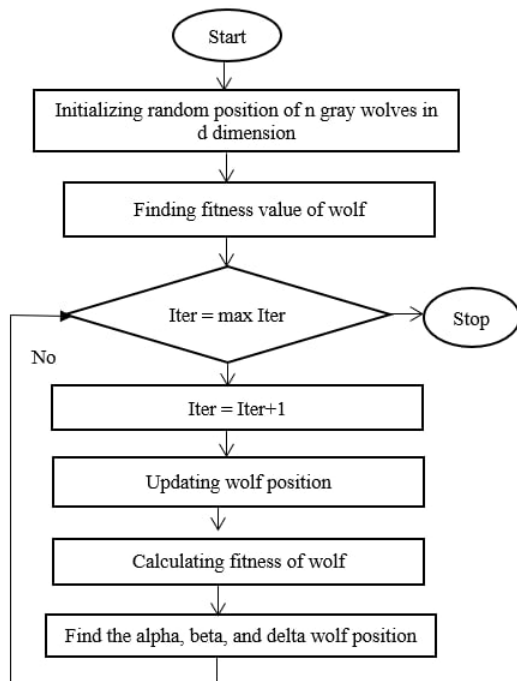
The wolves advance on the prey, making the victim's position the final position of Alpha.

$$|A| < 1$$

Pseudocode of GWO is described in Algorithm 2.

**Algorithm 2 GOW**

Initialization of grey wolf population  $X_i$  ( $i = 1, 2, \dots, n$ )  
 Initialize  $a$ ,  $A$ , and  $C$   
 Calculating the fitness of every one of the search agents  
 $X_\alpha =$  best search agent  
 $X_\beta =$  2<sup>nd</sup>-best search agent  
 $X_\delta =$  3<sup>rd</sup> best search agent While ( $t <$  max number of iterations)  
 For every one of the search agents  
 Updating position of current search agent through Eq. (16)  
 End For  
 Update  $a$ ,  $A$ , and  $C$   
 Calculating fitness of all of the search agents Updating  $X_\alpha$ ,  $X_\beta$ , and  $X_\delta$   
 $t = t + 1$  End while  
 Return  $X_\alpha$



**Figure 7.** Flow chart of the grey wolf optimization algorithm

**5. SIMULATION RESULTS AND DISCUSSIONS**

Figure 8 depicts the simulation model for PI-controlled BLDC motor drive. According to the requirements, a reference

speed is set. The feedback path feeds the observed speed to the comparator. The PI controller processes the error signal. The motor's various parameters are listed in Table 1. Various algorithms are used for evaluating the PI controller's settings (Kp, Ki). For various systems, the GWO algorithm is used for evaluating the PI controller parameters, and the output response has been put to comparison with the PSO-based function. The PSO and GWO based PI controller design for BLDC system can be represented by the block diagram in Figure 9.

**Table 1.** BLDC Motor parameters

Rated voltage	80 v
Rated speed	1500 rpm
Rotor Inertia - J [kg-m <sup>2</sup> ]	5.5e-3
Inertia, J (Kg-m <sup>2</sup> )	
Resistance - Rs	1.43ohm
Inductance - Lls	9.4e-3H
Friction Coefficient -B	2e-3
Torque constant	10N.m
Rotor Flux	0.2158 wb
Number of pole pairs (P)	4

Depending on the Integrated-Square Error (ISE) between C(s) and R(s) specified as objective function J in Eq. (19), the GWO or PSO block will be fed back for minimizing the ISE to achieve appropriate PI Controller settings, i.e., Kp, Ki of PI controller producing a satisfying response.

$$E(S) = R(S) - C(S)$$

$$J = \int_0^{\infty} e^2(t). dt \tag{19}$$

**Table 2.** Basic parameter values of the algorithm

Algorithmic parameters	PSO	GWO
Max. no. of search agents	50	40
Iterations	40	30
Dimension	2	2
Lower bound	[5 150]	[5 150]
Upper bound	[20 500]	[20 500]

Table 2 shows the number of search agents and iterations for both methods. Comparably, Table 3 shows the Ki and Kp values for the two approaches. The output responses for the two cases are produced with the use of controller parameters listed in Table 4, as it has been shown in Figure 10, which shows that in the case of a GWO-based PI-controlled system, peak overshoot and settling time are shorter compared to in PSO-based system. Table 4 compares the performance of the two systems in the case when the motor is set to 1500 rpm.

**Table 3.** PI controller parameters for different

Kp	Ki	Speed	algorithm
9.660415791715693	187.2824720285225	1454.4	PSO
9.9652	224.4844	1455.2	GWO
10.7253	238.4021	1458.1	PSO
11.8736	298.3538	1461	GWO
12.5275	179.7175	1460	PSO
4.229	193.3457	1463	GWO
15.8736	298.3538	1464	PSO
17.22086	246.6088	1467	GWO

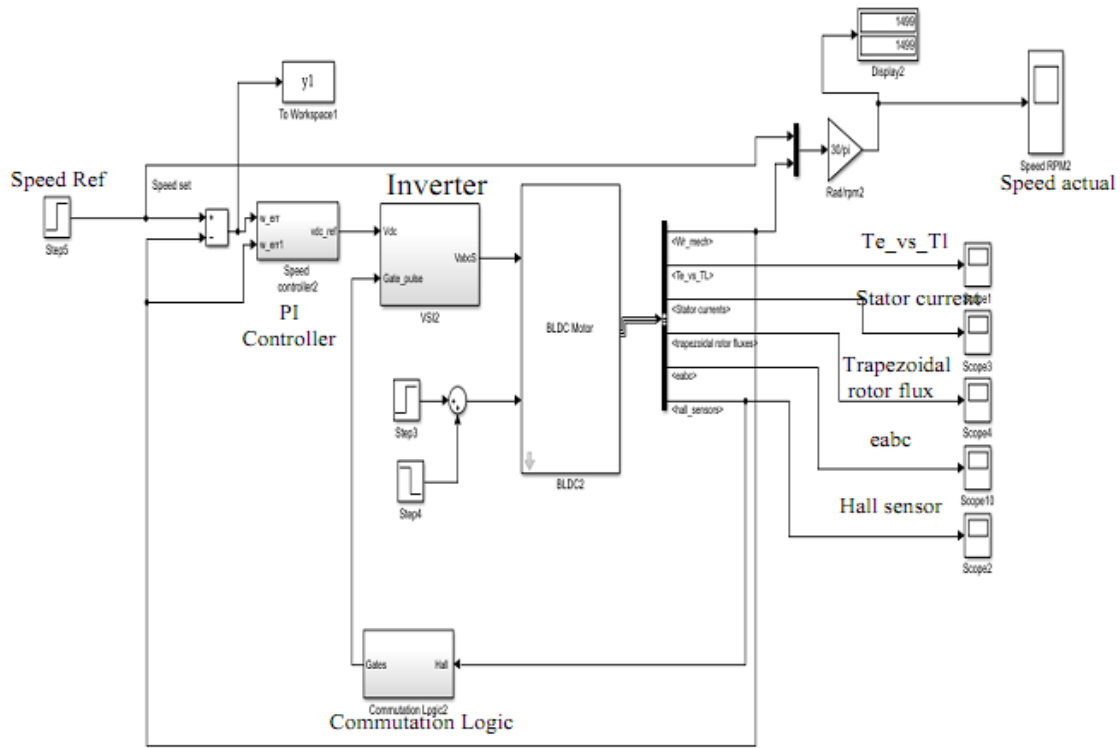


Figure 8. A BLDC Motor simulation model

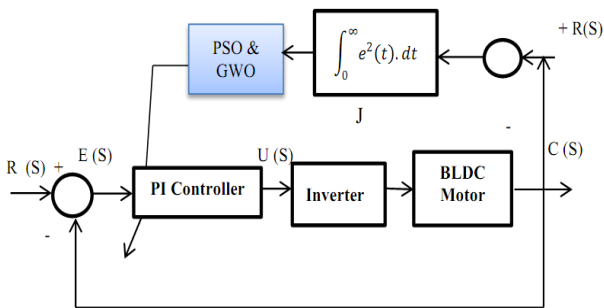


Figure 9. PSO&GWO-based PI Controller design

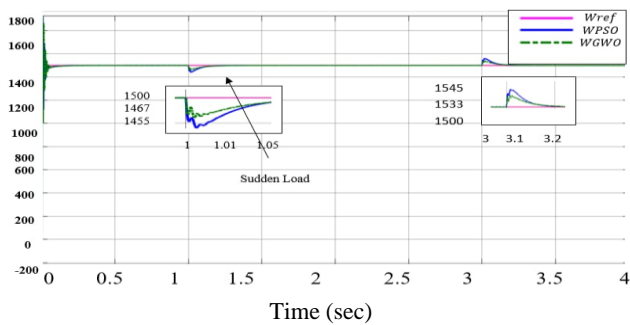


Figure 10. Speed response in 1500 rpm versus time under load 10N.m

Table 4. Performance of the PSO and GWO based PI controllers

Algorithm	Rise time (s)	Overshoot (s)	Settling time
PSO	3.012	3	2.3
GWO	3.015	2	2.25

The three-phase currents for ramp response are shown in Figure 11. The magnitude current is 5 A from 0 to 0.9 sec since

the motor is running at no load, and it increases to 10 while the motor is running at full load. Figure 12 show the three-phase back emf and Figure 13 Electromagnetic Torque developed in N-m.

When comparing this work with another work presented in 2018 [2], which is a comparison between NCS and CCS, it was found that the results of PSO and GWO are the best from the Figure 14 and 15.

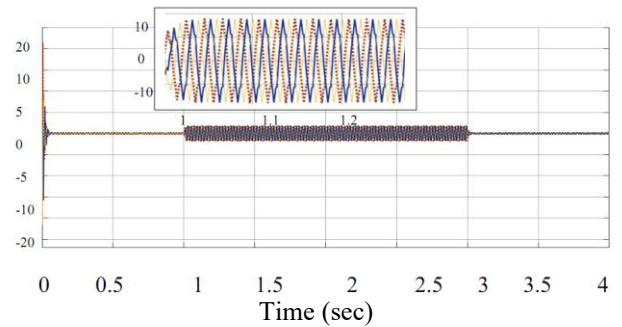


Figure 11. Three-phase stator current

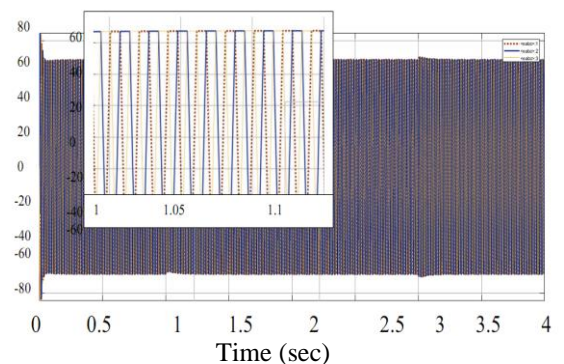


Figure 12. Three Phases back emf

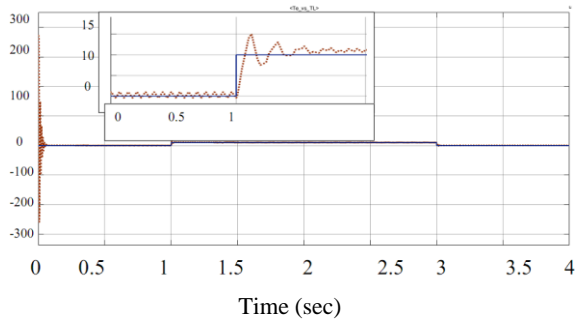


Figure 13. Electromagnetic Torque developed in N.m

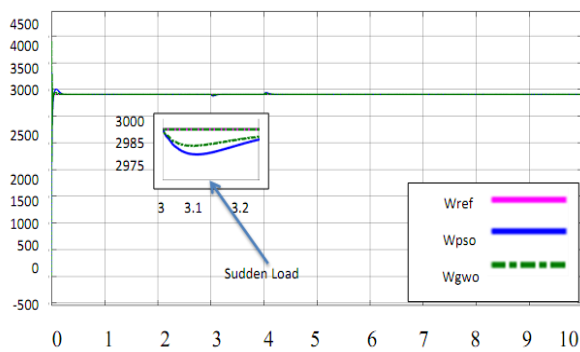


Figure 14. Speed response in 3000 rpm versus time

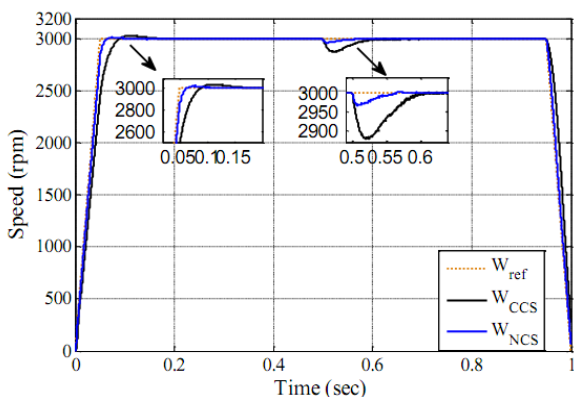


Figure 15. Speed response in 3000 rpm versus time

## 6. CONCLUSIONS

The GWO and PSO are proposed for use in designing an optimum PI controller for the BLDC motor speed regulation in this study. Both GWO and PSO were utilized for getting the optimal reaction. A PI controller was used for controlling the BLDC speed. BDC motor's speed is controlled using a new optimization approach (GWO) which surpasses the PSO method in this research. Yet, PSO has some drawbacks, such as being easy to slip into a local optimum in a high-dimensional environment and having a slow rate of convergence. Several types of research were undertaken in the previous few decades for improving PSO. The data suggest that the machine settles down in a shorter time than the PSO-based method. Even though the GWO approach takes somewhat longer to alter the PI controller parameters regarding a BLDC motor compared to the PSO approach, the significant improvements in time domain performance justify

its adoption. The suggested approach could give the field of BLDC motor drive system controller design a new dimension.

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