An Optimal Tilt Integral Derivative Applied to the Regulation of DC Link Voltage in a Stand-Alone Hybrid Energy System



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https://doi.org/10.18280/jesa.540410	ABSTRACT
Received: 25 January 2021	This paper presents an application of fractional control scheme named Tilt Integral
Accepted: 17 May 2021	photovoltaic (PV) system and a battery bank (BB). A three-level NPC inverter is inserted in order to increase the efficiency of the energy injected into the AC load. Variation in
battery bank, buck-boost converter, DC link voltage, genetic algorithm, photovoltaic system Proportional-Integral-Derivative	solar radiation or AC load may cause power imbalance, which leads to variation in DC link voltage. As a solution, a buck-boost converter is connected between the DC link and the battery bank to ensure the transfer of energy in both directions. The parameters of

(PID), Tilt-Integral-Derivative (TID)

TID controller were tuned using a powerful optimization technique known a Genetic Algorithm (GA) by minimizing the Mean Square Error (MSE) used as a performance index. The effectiveness of the proposed TID controller is demonstrated through a comparison with a conventional Proportional-Integral-Derivative (PID) controller, whose parameters are computed by the pidtool function of the Matlab/Simulink tool where the DC link voltage behavior is previously modeled by a capacitor transfer function. The obtained results show that the proposed TID controller provides a stable DC bus with low chattering, regardless of the rapid irradiation and load changes, when compared to a conventional PID controller.

1. INTRODUCTION

Because of the increased demand for energy consumption and the limited sources of fossil fuels in the world, the effective use of Renewable Energy (RE) is becoming more and more important. Furthermore, the increasing cost of limited sources and the difficulty to extract it along with the challenge of climate change have raised the need for renewable energy and have accelerated the transition to cleaner, more efficient electric power methods. There are several types of RE such as wind, solar, biomass and geothermal and each has its own merits and demerits [1-4].

Solar energy is the light and heat emitted from the sun converted to current by a device called a photovoltaic panel, where the photovoltaic cascades conversion fall into two categories: grid-connected systems and stand-alone. The latter configurations are applied in low power systems and require the battery bank to store the PV energy.

The power delivered by the PV system depend on two parameters: irradiance solar and ambient temperature [1-3]. Among the most famous applications, there is the autonomous system or called off-grid or Stand-alone abbreviated by "SPVS" used in several applications such as street lighting systems, remote areas and mobile military equipment for the desert and

border-outpost [4].

Depending on the PV energy requirements for supplying the load, sizing is required for the overall PV system. So, to convert the maximum power of a PV system to the AC load, voltage must be adapted to it under changing solar irradiation. So, the connection of power electronic devices between the PV system and the battery in addition to the AC load plays an important role in achieving balance [5-7].

Researchers use different technical controls for DC-link voltage. For stand-alone mode, Samrat et al. have used a simple proportional-integral (PI) to control DC-link voltage [8-10]. However, Benlahbib et al. [11] achieved the control of the DC voltage link in a hybrid energy system composed of a turbine and batteries by a fractional order PID (FOPID), where the turbine is based on a permanent magnet synchronous generator (PMSG). In another research, control techniques such as sliding mode control was used [10]. In grid connected systems, a particle swarm optimization (PSO) was used to set the PI parameters [12]. In contrast, the authors in this paper used for the first time an optimal TID controller to regulate the DC link voltage in a DC-DC buck-boost converter connected to a lead acid battery bank. its parameters were determined using a meta-heuristic optimization technique such as a Genetic Algorithm (GA).

This paper uses the performance of the hybrid PV/battery system to satisfy the load requirements under any meteorological condition. The PV system and the batteries bank are coupled to the DC Bus through boost and reversible Buck-Boost converters respectively. A three phase three-level inverter (NPC) is the static converter used to feed the load. The proposed PV/Battery block is shown in Figure 1.

The first part in this paper presents an algorithm of Maximum Power Point tracking (MPPT) called the perturbation and observation (P&O) method applied to the PV system to extract maximum power by increasing or decreasing the duty cycle D through a Boost converter.

Secondly, a control technique applied is to charging/discharging the battery through a Buck-Boost Converter (BBC) to keep the DC link voltage stable during variations in load demand or source power. By controlling the DC link voltage, this control technique ensures the power balance in the system. A DC link voltage connected with DC-DC converters interface the PV panels with the battery [13, 14].

Finally, as we mentioned previously, the three-level inverter NPC is used to feed the AC load by a LC filter (Figure 1), which serves to reduce the high-frequency harmonics [14, 15].

The proposed PV/battery hybrid system was modeled, managed, simulated and validated under Matlab/Simulink tool. Furthermore, the results obtained are listed to verify the effectiveness under variable meteorological conditions [8]. Thus, the authors propose an optimal TID control to regulate the DC-link voltage through a two-quadrant buck-boost DC-DC converter integrated in the proposed system mounted exactly with the battery. Here, the TID control takes the variations of solar radiation and load into account to ensure robustness.



Figure 1. The proposed block PV/Battery

2. MODELING AND CONTROL OF THE PV SYSTEM



Figure 2. Equivalent circuit of the PV module

In general, the PV cell is often made of a semiconductor material such as silicon, germanium, gallium, arsenide, etc. It can therefore absorb the light energy and convert it into electrical current, where its characteristics are varied according to two weather conditions: solar radiation and absolute temperature. Indeed, in the modeling step of actual PV cell behavior, several PV models have been proposed [9, 10]. Among them, the equivalent electrical circuit, shown in Figure 2, is the most widely one used for computing the predicted current model I_{pv} [16-19].

Two resistances and $\stackrel{\cdot}{a}N_P$ diode is available. The shunt resistance $\frac{N_S}{N_P}R_{sh}$ illustrates the loss, which is a tiny leakage current that flows through the parallel part (Order of $k\Omega),$ the series resistance connection $\frac{N_S}{N_P}R_s$ (approximately 1 Ω) [20]. Number of cells in series is denoted N_s and in parallel N_p , the relationship between the output current I_{pv} and the voltage V_{pv} is written as [4, 21-23]:

$$I_{pv} = N_P I_{ph} - N_P I_s \left(e^{\frac{V_{pv} + R_S}{N_S + N_P} I_{pv}} e^{\frac{K \cdot T}{q}} - 1 \right)$$

$$- \frac{N_p V_{pv} + R_S N_S I_{pv}}{N_S R_p}$$

$$(1)$$

where:

The generated photocurrent I_{ph} and Saturation current I_s are given respectively by:

$$I_{ph} = \frac{s}{1000} \left(I_{sc} + K_i (T - 295.15) \right)$$
(2)

$$I_{s} = I_{RS} \left[\frac{T}{295.15} \right] \left(e^{\frac{qE_{g}}{A.K} \left(\frac{1}{295.15} - \frac{1}{T} \right)} \right)$$
(3)

where: $K=1.38 \times 10^{-23}$ j/K is Boltzmann constant.

The parameters of PV cell are summarized in the Table 1.

Table 1. PV cell parameters



Figure 3. Characteristics of Sanyo HIB-225HDE1 under varying irradiation

PV Voltage (V)

20

60

40

PV Voltage (V)

The Eqns. (1) to (3) show that the current generated by the PV array depends on solar irradiation.

Figure 3 shows power and current characteristics of the Sanyo HIB-225HDE1 under effect solar irradiance.

P&O is the first algorithm among the MPPT algorithms used in PV systems. To implement it in direct mode control, a duty cycle D and a sampling frequency are assigned at the beginning as illustrated in Figure 4 [8, 17, 18]:



Figure 4. P&O algorithm

$$\begin{cases} \Delta V_{pv} = V_{pv}(k) - V_{pv}(k-1) \\ \Delta I_{pv} = I_{pv}(k) - I_{pv}(k-1) \text{ and} \\ \Delta P_{pv} = P_{pv}(k) - P_{pv}(k-1) \\ \frac{\Delta P_{pv}}{\Delta V_{pv}} = \begin{cases} = 0, at \ MPOP \\ > 0, left \ MPOP \\ < 0, right \ MPOP \end{cases}$$
(4)

3. THREE-LEVEL INVERTER MODELING AND CONTROL



Figure 5. Three phase three-level inverter NPC

Generally, the Three-Level Inverter NPC is attached between the DC link voltage and the load of the system. It used to regulate the voltage and frequency of the charge. Note here that our system is an off-grid, so the load voltage must be regulated in terms of frequency and voltage amplitude [8, 9]. The topology of a three-level inverter is illustrated in Figure 5, where the upper switches of the three-level inverter are S_{k1} , S_{k2} in the ON state, which corresponds to the state V_{dc1}' . If S_{k3} , S_{k4} is on the lower switches, which corresponds to state'- V_{dc2}' . This results in condition '0' when the auxiliary switches are on S_{k3} , S_{k4} [15].

The switch connection function F_{KS} indicates the opened or closed state of the switch S_{KS} , $F_{ks} = 1$ when S_{ks} close and $F_{ks} = 0$ when S_{ks} open (s=1, 2, 3, 4 and k=1, 2, 3)

The functions F_{km}^{b} of connection are given by:

$$\begin{cases} F_{k1}^{b} = F_{k1}.F_{k2} \\ F_{k0}^{b} = F_{k3}.F_{k4} \end{cases}$$
(5)

whereas:

m = 1: The upper part of the NPC inverter works. m = 0: The lower part of the NPC inverter works. Leg voltage V_{A0}, V_{B0} and V_{C0} can be written as:

$$\begin{cases} V_{A0} = F_{11}^{b} V_{dc1} - F_{10}^{b} V_{dc2} \\ V_{b0} = F_{21}^{b} V_{dc1} - F_{20}^{b} V_{dc2} \\ V_{c0} = F_{31}^{b} V_{dc1} - F_{30}^{b} V_{dc2} \end{cases}$$
(6)

Simple output voltages are written as:

$$\begin{bmatrix} V_A \\ V_B \\ V_C \end{bmatrix} = \frac{1}{3} \begin{bmatrix} 2 & -1 & -1 \\ -1 & 2 & -1 \\ -1 & -1 & 2 \end{bmatrix} \left\{ \begin{bmatrix} F_{11} \\ F_{21} \\ F_{31} \end{bmatrix} V_{dc1} - \begin{bmatrix} F_{10} \\ F_{20} \\ F_{30} \end{bmatrix} V_{dc2} \right\}$$
(7)

As AC loads are often single-phase in nature in the delivery system, the current would not be the same in various phases, and would not be equal in each phase because of the unbalanced load voltage attached to the inverter NPC. This unbalanced voltage drops at the level of the LC filters used at each phase, where the acceptable limit for the voltage unbalance factor is <1%. To compensate for this voltage unbalance at the AC bus, the error between the Root Mean Square (RMS) of the voltages and the reference voltage is fed into the PI controller. The output of the PI controller is multiplied by а unit sine wave generator $(V_{a-gen}, V_{b-gen}, V_{c-gen})$. Using V_{a-ref}, V_{b-ref} and V_{c-ref} , Pulse width Modulation (PWM) pulses are produced for the three-level NPC inverter to be turned on/off [24].

The function of the control scheme (Figure 6) is to separate modulation indexes for three phases in order to balance the unbalanced voltages on the AC bus [24, 25].



Figure 6. PWM of a three-phase three -level inverter

4. MODELING AND CONTROL OF BATTERY

4.1 Battery modeling

Out of all models available, the analogous circuit model is the most widely used for dynamic simulation. The battery has a high-energy capacity and can supply power at almost constant voltage if the cycles are properly regulated for charge/discharge. A lead-acid battery is utilized in this work, since it is more convenient for green systems due to its low cost and availability [13].

In our proposed system, the modeling of the battery was done with a constant resistive value of the controlled series attached voltage source shown in Figure 7 formulated by [8, 24]:

$$\begin{cases} E = E_0 - K \frac{Q}{Q - \int idt} + Aexp\left(-B \int idt\right) \\ V_{Battery} = E - R_{in}I_{Battery} \end{cases}$$
(8)

where, the parameters are summarized in the Table 2.

Figure 7 shows the circuit of the implemented model of the battery in Matlab/Simulink tool [8, 9].

Table 2. Battery parameters

Parameter	Denomination	Unit
E ₀	Constant voltage	V
Q	Battery capacity	Ah
K	Polarization voltage	V
А	Exponential voltage	V
В	Exponential capacity	Ah ⁻¹
V _{Batterv}	Battery voltage	V
Rin	Internal resistance	Ω
∫idt	Extracted capacity	Ah
I _{Battery}	Battery current	Α



Figure 7. Non-linear typical battery model



Figure 8. The proposed block PV/Battery

4.2 Battery control strategy

In order to obtain the required power, it is essential to regulate the current generated by the battery. One must bear in mind the constraints related to the existing charge and discharge and the overall state of charge (SOC) limits. The following criteria must be reviewed by the SOC:

$$SOC_{min} \le SOC(t) \le SOC_{max}$$
 (9)

where, the maximum and minimum acceptable storage capacities are respectively, SOC_{min} and SOC_{max} .

The battery has an important role as an intermediate between the PV and the load. Generally, it has two operating states, namely: charging and discharging. To control the DC voltage, the BBC controller has two switches namely Q_1 and Q_2 , where the working principle is detailed as follow:

- Charging mode: the converter functions as a buck circuit by the switch Q₁ when the voltage on the DC link is greater than the reference voltage;
- Discharging mode: the converter functions as a boost circuit by closing the switch Q₂ when the voltage on the DC link is lower than the reference voltage.

Whatever the constraints or load variations, it is still held at a steady voltage [15].

The proposed PV/Battery block is coupled with the threelevel inverter to feed the three phase loads by a DC link voltage as shown in Figure 8 [8, 9, 24].

The BBC is used to maintain the value of the DC link voltage as desired and to ensure a charging or discharging of the battery according to the PV system (Figure 8). The block diagram of the controller is detailed in the same Figure 8. By using the controller in the simulation under Matlab/Simulink tool (detailed in section 5), the batteries must be connected in series to have a voltage lower than the reference DC link voltage (V_{dc}). In our proposed system the battrey voltage is fixed at 300V; this allows to choose $V_{dc} > 300V$ (our choice is $V_{dc} = 700V$). In this analysis, the battery discharge depth is considered to be 60%. Furthermore, it is expected that electrical power should be supplied to the 8 kW load.

Generally, the battery serves either as a load or as a power source, depending whether there is surplus or a loss of power. It can charge or discharge under defined limits according to weather conditions. In this analysis, due to high power conditions, surplus power is initially supplied to the battery bank before it exceeds its upper limit of storage capacity [26].

The flow chart in Figure 9 shows the above control coordination between the PV and battery, where the battery's lower and upper SOC limits are held at 0.2 and 0.6, respectively [24].

4.3 TID Controller

The TID corrector is a control system that improves the performance of a closed-loop system or process. The structure of the TID $(K_T, K_I, K_D \text{ and } N)$ is similar to that of PID $(K_P, K_I, \text{ and } K_D)$, except that the proportional behavior is replaced by an inclined proportional behavior having a transfer function $S^{-\frac{1}{N}}$. The inclined behavior provides a feedback gain according to the frequency, which is inclined relative to the gain / frequency of the conventional compensator. Thus, the whole compensator is called TID. A representation of a functional diagram of TID control is shown in Figure 10 [27-31].



Figure 9. DC link voltage control algorithm



Figure 10. TID controller bloc diagram

In Figure 10, the transfer function of the TID controller expressed in parallel form is given by:

$$K(S,X) = \frac{K_T}{S^{\frac{1}{N}}} + \frac{K_I}{S} + K_D \cdot S$$
(10)

4.4 GA Optimization

The desired TID controller has four variables, which are:

$$X = (K_T, N, K_I, K_D)^T$$
(11)

The optimal vector X* is determined from minimizing the Mean Square Error (MSE) given by:

$$J = \frac{1}{n} \sum_{i=1}^{n} \left(V_{dC}(i) - V_{dC_simu}(i) \right)^2$$
(12)

where, V_{dC} is the reference desired DC link voltage, V_{dC_simu} is the output DC link voltage given by simulation under Matlab/Simulink using GA algorithm generated by control parameters for n sampled points. This optimization problem is solved under the following constraints:

$$\begin{bmatrix} 0\\2\\0\\0\\x_{min} \end{bmatrix} \le \begin{bmatrix} K_T\\N\\K_I\\K_D \end{bmatrix} \le \begin{bmatrix} inf\\3\\inf\\inf\\inf\\x_{max} \end{bmatrix}$$
(13)

The block diagram (see Figure 11) shows the controller optimization strategy:



Figure 11. Flowchart describing the tuning parameters of the TID controller by the GA

According to Figure 8, the discrepancy voltage, generated from comparing the two voltages V_{dC} and V_{dC_simu} is used to formulate the fitness function, given by Eq. (12). It is then transferred to the MATLAB software's based on GA function, where its setting parameters are previously selected by the designer using some existing guidelines [32-34]. Then, the GA algorithm generates randomly initial populations where the fitness function is evaluated in each one. This yields also to providing the first optimal parameters of the TID controller. The three steps such as reproduction, crossing and mutation of populations are performed in order to find other optimal TID parameters better than those existing in the preceding populations. Next, a judgment test is performed to select the best solution, in which a proper attenuation of the fitness function is done. These steps are repeated as the stopping criterion is not reached. Finally, the GA is achieved by providing the best optimal solution that allowing to determine the TID transfer function. As a result, according to Figure 11, the GA step-procedures are summarized as follows:

initialization of population;

While the goal or the number of generations not obtained Crossover step:

perform crossover between the bests parent chromosomes; Mutation step:

evaluate fitness offspring after choose mutation points and perform mutation

Reproduction step: create new population End

Table 3 summarizes the tuning parameters of the GA function under Matlab/Simulink, which are the same used in [34].

Parameter		Value
Programme execution		20
Population size		100
Generation number		20
Reproduction	Elite count	2
	Crossover	0.8
Mutation function		Constrain dependent
		Scattered
Crossover function		Forward
Migration	Direction	0.2
Ū.	Fraction	

5. ANALYSIS OF SIMULATION RESULTS

In this section, the control loop is implanted in Matlab/Simulink software using the sampling time $T_s = 10^{-5}$ S; where the numerical values of the PV and the battery systems are summarized in Tables 4 and 5, respectively.

Table 4. PV Array parameters

Components	Ratings values
Module Type	Sanyo HIP-225HDE1
Number of Cells	60
Series module	06
Parallel module	06
Voc, Isc, Vmmp,	41.79V, 7.13A, 33.9V, 6.63A
Immp	
Rs, Rp, Isat, Iph, Qd	00.204Ω,1830.7Ω,3.0815e-07A,
	7.145A,1.6

The optimal values obtained of the TID are $K_T = 21.0521$, N = 2.0089, $K_I = 3.8001$ and $K_D = 0.0299$ as shown in the fitness plots provided by the algorithm during the extraction process (Figure 12). The pidtool function is applied to provide the three PID gains $K_P = 1.042$, $K_I = 355.5$ and $K_D = 0$.

Table 5. Battery parameters

Components	Ratings values
Battery type	Lead-Acid
Nominal voltage	300 V
Capacity rating	6.5 Ah
Initial state of charge	60%

The robustness test was performed as shown in Figure 13: - nominal temperature value T=25°C;

- rapidly change in solar irradiance 500 W/m² to 1000 W/m²;
- rapidly change in AC load power 6KW to 8KW.



Figure 12. The best obtained fitness curve through GA algorithm

The robustness of the proposed controller was assessed under DC link fixed at 700 V during variation of irradiance and load. For this study, the response of the TID and PID controllers were taken into consideration. The DC link voltage shown in Figure 14 is nearly constant, but with some negligible variation due to the large delay between PV and load changes recovered by the battery as shown in Figure 15.



Figure 13. Irradiance and power load variations



Figure 14. DC link voltage of the PV/Battery system using TID and PID controllers



Figure 15. Powers distribution curve of PV/battery and load

According to Figure 14, one can clearly observe that the DC link voltage responses and the resulting control given by the TID controller are better than those provided by the PID controller. The superiority of the TID controller manifests itself in smooth DC link voltage response, less sensitivity against load power variations and elimination of the steadystate electrical power oscillation. In addition, it can be seen from Figure 15 that the BBC controller's output is very satisfactory because the battery bank power changes (charges/discharges) to maintain the balance of the system power under various irradiance and load conditions. In reality, if the AC load power is greater than the produced PV power, the BBC controller is programmed to discharge the power of battery into the AC load. In addition, the controller is also capable of charging the battery power when the load power is lower than the produced PV power.

The peak power of the battery (Figure 15) in transitional regime is justified by a delay in the power supplied by the PV due to the influence of the MPPT algorithm, which depends on voltage and current to provide a maximum power. The SOC, as shown in Figure 16, is at 60% at the beginning, and then it decreased in transitional regime until 0.2 S before increasing until 1.25 S (charging period). Next, it is discharged up to 5 S (the case of increasing power load and decreasing PV power). Finally, it is charged again when the power load decreases and the PV power increases.

From the above results, it is easy to observe the effectiveness of:

- The proposed control of the energy variation as shown in the flow chart of Figure 9;
- The optimal TID controller for our proposed PV/Battery system.



Figure 16. State of Charge SOC of the battery



Figure 17. The output phase current and inverter voltage



Figure 18. Zoom of the output phase current and inverter voltage



Figure 19. Spectrum analysis of current and voltage of the inverter

The current and voltage waveforms of the three levelinverter of the load when TID control is implemented are shown in Figure 17. There is also a zoom in on a small range as shown in Figure 18.

This result demonstrates the efficacy of the suggested threelevel inverter control and the better quality of the supplied power. To confirm that the two singles current and voltage produced by the NPC inverter for the proposed control are good, a Total Harmonic Distortion (THD) calculated (must be < 5%), were 1.74% and 1.49% (see Figure 19).

6. CONCLUSION

A comparative study based on the performance and robustness of the regulation of a DC link voltage in a hybrid system composed of a PV/battery system using PID and TID controllers has been presented in this paper. It was assessed for different changes in solar irradiance and power load. The TID controller parameters tuned by GA algorithm exhibit better performance and robustness than those offered by a PID controller. The optimal PID control depends on the transfer function by using *pidtool function* in Matlab/Simulink tool. The simulation results confirm that the TID control allows rapid tracking of the desired DC link voltage under various tests, also the oscillation problem is significantly reduced.

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NOMENCLATURE

BB	battery bank
BBC	Buck-Boost Converter
GA	Genetic Algorithm
MPPT	Maximum Power Point tracking

MSE	Mean Square Error
NPC	Neutral Point Clamped
P&O	perturbation and observation
PID	Proportional-Integral-Derivative
PWM	Pulse width Modulation
RE	Renewable Energy
SOC	state of charge
TID	Tilt Integral Derivative
THD	Total Harmonic Distortion

Greek symbols

I_{pv}	Photovoltaic current, A
I _{ph}	Photocurrent, A
I _s	Saturation current, A
V _{pv}	Cell output voltage, V
Â	Quality factor.
q	Charge of electron, C
Т	Cell temperature,
K	Boltzmann constant, 1.38×10^{-23} j/K
S	Solar irradiance, W/m ²
K _i	Current coefficient,
E ₀	Constant voltage of battery, V
Q	Battery capacity, Ah
V _{Battery}	Battery voltage, V
I _{Battery}	Battery current, A
В	Exponential capacity, Ah ⁻¹
V _{dc}	DC link voltage, V
R _{in}	Internal resistance, Ω
\mathbf{Q}_1	Switch 1,
Q_2	Switch 2,
∫idt	Extracted capacity, Ah